

## SCM1212B 3.3 -5V Transformer Driver for Isolated Power Supplies

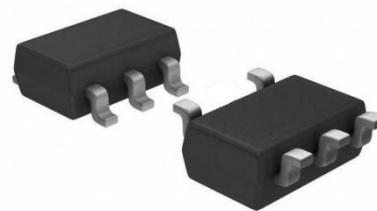
### Features

- Push-Pull driver controller
- 2.7-5.5V wide input voltage
- Low conduction resistance 200mΩ
- Two built-in power MOSFETs with high symmetry ,quasi-complementary driver
- Limited MOSFET's current while turning on the power
- Over current protection and Short circuit protection
- Over temperature protection
- Under voltage protection
- Increase the dead time under light load
- Meet AEC-Q100 automotive standard

### Application

- Isolated Interface Power Supply for CAN, RS-485, RS-232 etc.

### Packaging

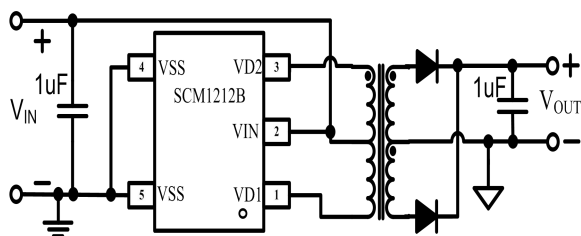


Optional Packaging of Product:SOT23-5 please refer to "Order Information" for details of silk screen.

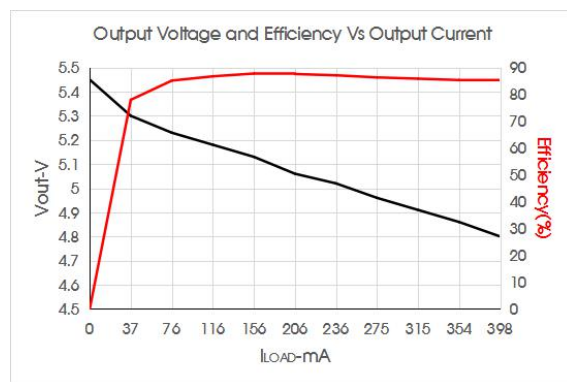
### Description

The SCM1212B is a driver integrated with MOSFETs for the primary side of a push-pull power. The chip can function at the low input voltage of 3V, and will not be damaged under the high input voltage of 9V in one second. The device contains two drive systems and two power MOSFET, and each drive system decides circuit's direction of one primary winding. Each drive system includes one power NMOSFET. Two drive systems operate in turns to achieve push-pull drive. Two power NMOSFETs are highly symmetrical in order to decrease the degree of magnetic biasing of push-pull topology. The SCM1212B is also integrated with three key technologies to improve reliability. The first one is soft start function, which limits the current of MOSFET to avoid the damage caused by the large current impact when turning on the power, and quantitatively guarantees the capacitive load in CC load mode. The second one is output over-current and short circuit protection. On the one hand, the SCM1212B distinguishes between the over-current protection and the short-circuit protection, which can control the short-circuit temperature rise while guarantee the capacitive load. On the other hand, the chip can adjust protection's threshold based on the input voltage and temperature to improve the consistency. The last one is over temperature protection, when the chip operates beyond the specified temperature, it will enter the dormancy state automatically. It will not operate until the temperature decreases to the set point.

### Typical Application Circuit



### Function Curves



Contents

Features.....1

Packaging.....1

Application.....1

Function Description.....1

Typical Application Circuit.....1

Function Curves.....1

Pin Packaging.....2

Inter Block Diagram.....2

Pin description.....2

Thermal Resistance Information.....3

Absolute Maximum Ratings.....3

Recommended Operating Conditions.....3

Electrical Characteristics .....3

Typical Curves.....4

Parameter Measurement Information.....5

Product Operation Mode .....6

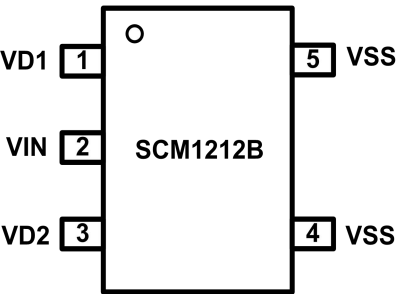
Extended Output Design .....8

Application Circuit.....8

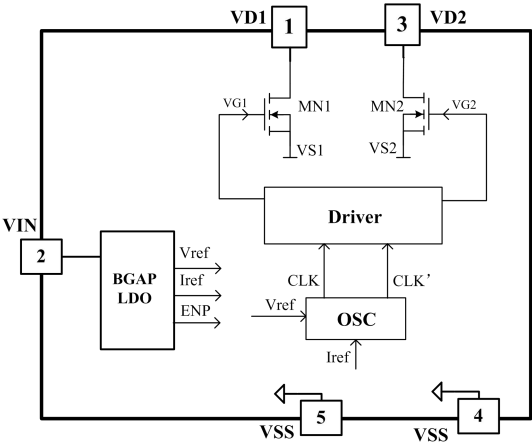
Suggestions of Power Use.....8

Order,Packaging and Packing.....9

Pin Configuration



Inter Block Diagram



Pin Description

No.	Name	I/O	Description
1	VD1	I	Built-in power LDMOSFET's drain.It becomes soft drive when starting or detecting short circuit, limiting the current flowing through LDMOS tube. If $T_{D\_OSP}$ continuously detects a short circuit, the chip will enter the dormancy state. The dormancy time is $T_{sleep}$ , and the chip will restart again after the dormancy.
3	VD2	I	
2	VIN	P	Supply voltage input
4	VSS	P	Device ground. Connect this pin to board ground
5	VSS	P	Power ground

Absolute Maximum Ratings

General test conditions:Free-air,normal operating temperature range(unless otherwise specified).

Parameters		Min	Max	Unit
Input Voltage	V <sub>VIN</sub>	-0.4	9	V
Drain Voltage of MOSFET	V <sub>VD1</sub> /V <sub>VD2</sub>	-0.7	26	V
Operation Junction Temperature Range	T <sub>J</sub>	-40	150	°C
Storage Temperature	T <sub>STG</sub>	-55	150	
Soldering Temperature (Allowable reflow soldering temperature of chip within 10 seconds)			260	
Moisture Sensitivity Level	MSL	MSL3		
Rated Value of ESD	HBM		5000	V
	CDM		1000	

Note:if the value exceeds the stress value listed in the table's "Absolute Maximum Ratings",it may cause permanent damage to the components.If the product operates in the maximum rated condition for a long time,the reliability of the components may be affected.All voltage values take GND as basis reference.The current refers to the current between positive input and negative output of the specified terminal.

# Recommended Operating Conditions

Unless otherwise specified,the following parameters are measured in the conditions of  $V_{VIN}=5V$ .

Parameters		Min	Max	Unit
Input Voltage	$V_{VIN}$	2.5	6	V
Drain Voltage of MOSFET	$V_{VD1}/V_{VD2}$	-0.7	18	V
Output Switching Current of Primary Winding	$I_{D1}, I_{D2}$	50	600	mA
Operation Junction Temperature	$T_J$	-40	150	°C

# Electrical Characteristics

Unless otherwise specified,  $V_{VIN}=5V$  and the environment temperature is 25°C.

Symbol	Corresponding Parameters	Test Conditions	Min	Typ	Max	Unit
Supply Section (VIN Pin)						
V <sub>VIN</sub>	Operating Voltage range		2.5		6	V
I <sub>RUN</sub>	Operating Current of Chip	V <sub>VIN</sub> =5V,No connection to VD1 and VD2		1.5	2	mA
I <sub>START</sub>	I <sub>VIN</sub> when V <sub>VIN</sub> is in under-voltage lockout	V <sub>VIN</sub> =2V		200		uA
V <sub>VIN_ON</sub>	Start-up Voltage	V <sub>VIN</sub> voltage increasing		2.5	2.7	V
V <sub>VIN_OFF</sub>	Voltage when V <sub>VIN</sub> is in under-voltage lockout	V <sub>VIN</sub> voltage decreasing	2	2.2		V
T <sub>OTP</sub>	Temperature of Over-temperature Protection	Environment temperature increasing		165		°C
T <sub>OTPH</sub>	Return Difference of Over-temperature Protection	Environment temperature decreasing		24		°C
Drain Port of MOSFET (VD1/VD2 Pin)						
B <sub>VDSS</sub>	Breakdown Voltage of MOS Transistor	V <sub>GATE</sub> =0V, I <sub>DS</sub> =100uA	26	35		V
R <sub>DS_ON</sub>	NMOS On Resistance	V <sub>VIN</sub> =3V,I <sub>DS</sub> =0.5A		200	500	mΩ
		V <sub>VIN</sub> =5V,I <sub>DS</sub> =0.5A		150	400	
I <sub>SOFT</sub>	Current of Soft Start	V <sub>VIN</sub> =3V,T <sub>J</sub> =25°C,V <sub>VD1</sub> =V <sub>VD2</sub> =3V		1000		mA
		V <sub>VIN</sub> =5V,T <sub>J</sub> =25°C,V <sub>VD1</sub> =V <sub>VD2</sub> =3V	600	800		
Switch Characteristics						
F <sub>OSC</sub>	Operating Frequency	No connection to VD1 and VD2	195	220	245	kHz
T <sub>DEAD</sub>	Maximum dead time	VD1,VD2 is in series with 200 Ω /1W power resistance ,to VIN,V <sub>VIN</sub> =5V		210		ns
	Minimum dead time	VD1,VD2 is in series with 50 Ω /1W power resistance to VIN,V <sub>VIN</sub> =5V		120		ns
T <sub>D_OSP</sub>	Delay Time of Short Circuit Protection	F <sub>OSC</sub> =220kHz	28	150		ms
T <sub>SLEEP</sub>	Sleep Time of Short Circuit Protection	F <sub>OSC</sub> =220kHz		1.5		s

## Typical Performance Curves

Unless otherwise specified, the following typical characteristic curves are obtained in the conditions of  $V_{IN}=5V$  and  $T=25^{\circ}C$ . Typical performance curves are obtained by testing the test circuit shown in Figure 5.

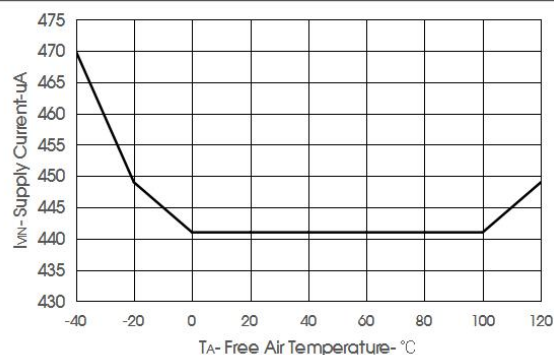


Figure 1 Supply average current of VIN VS Free-air temperature

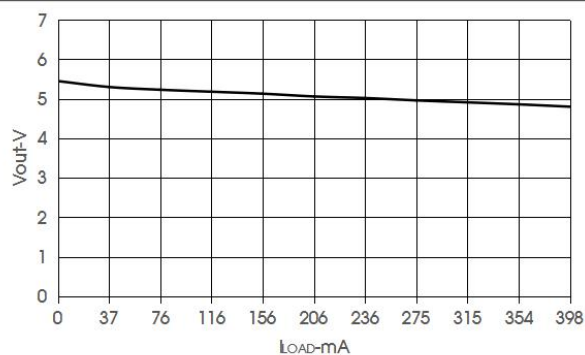


Figure 2 Output voltage VS Load current

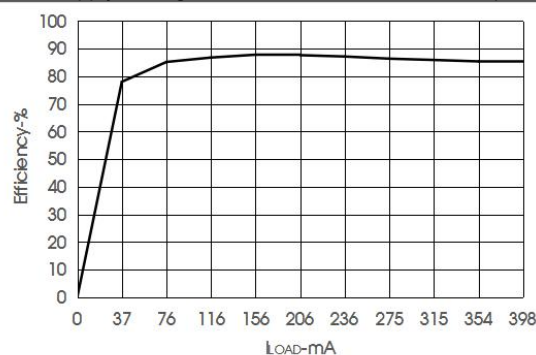


图 3 Efficiency VS Load Current

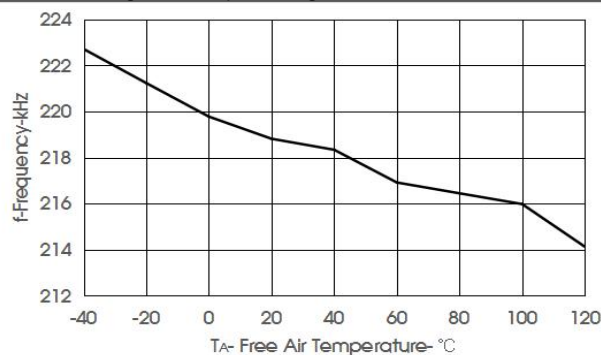


图 4 Switching Frequency of MOS Transistor VS Free-air Temperature

## Parameter Measurement Information

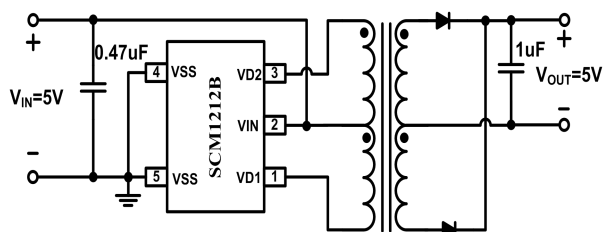


Figure 5 Schematic Diagram of Test Circuit for Function Curve

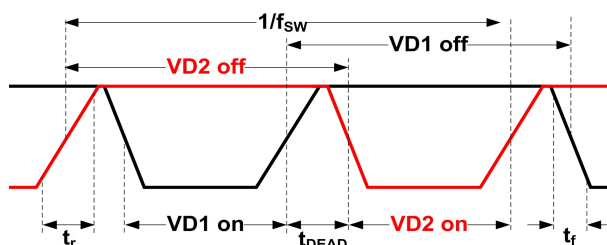


Figure 6 Circuit Sequence Diagram

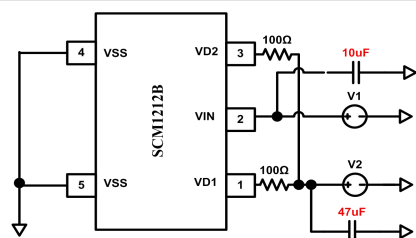


Figure 7 Schematic Diagram of Test Circuit for Switch Characteristics

## (1) Push-pull Converter

As shown in Figure 8 and Figure 9, the push-pull converter is a transformer with center tap, which can achieve the transmission of energy from the primary winding to secondary winding.

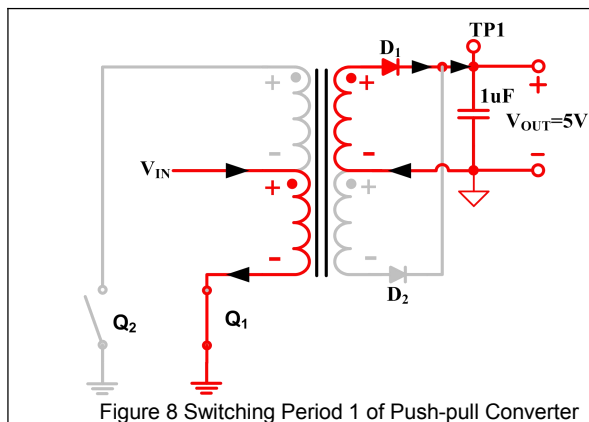


Figure 8 Switching Period 1 of Push-pull Converter

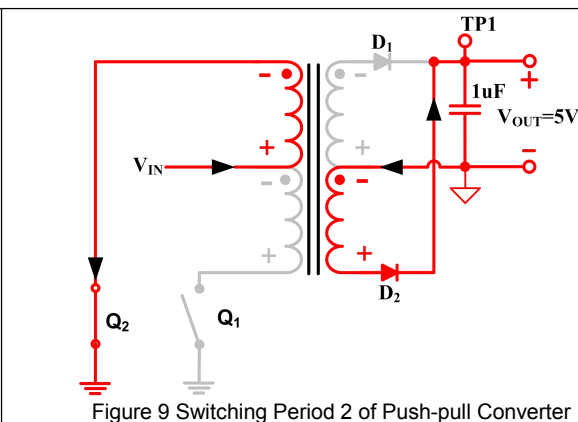


Figure 9 Switching Period 2 of Push-pull Converter

The drive waveform of drains  $V_{D1}$  and  $V_{D2}$  of two MOSFETs  $Q_1$  and  $Q_2$  are shown in Figure 6. Two MOS transistors are on alternatively and the times of the breakover periods of two transistors are equal, and there is a short period  $t_{BBM}$  between the two breakover periods that the two power transistors are not on. That is to say, the drive levels of two MOS transistors are quasi complementary in time sequence, that is the other transistor is off when one transistor is on, but there is a short period of dead time during the switch to prevent the backward current flow when two the transistors are not on simultaneously. As shown in red highlighted parts in Figure 8, when  $Q_1$  is on, input voltage  $V_{IN}$  drives a current which arrives at the reference ground through the lower half of primary winding of transformer and  $Q_1$ , and at the same time, the induced electromotive force of side winding charges output capacitor through diode  $D_1$ . Similarly, as shown in Figure 9, when  $Q_2$  is on, the induced electromotive force charges output capacitor through diode  $D_2$ . As continuously repeating the above process, the secondary winding of power converter obtain the needed power supply.

## (2) Magnetization of Magnetic Core

Figure 10 is the ideal magnetization curve of push-pull converter, and the vertical axis represents magnetic flux density  $B$  and the horizontal axis represents magnetic field intensity  $H$ . When  $Q_1$  is on, the magnetic flow is pushed to point  $A'$  from point  $A$ . Similarly, when  $Q_2$  is on, the magnetic flow is then pulled back to point  $A$  from point  $A'$ . The magnetic flux density  $B$  is proportional to the product of voltage of primary winding  $V_{LP}$  and breakover time of MOS transistors  $t_{ON}$ , which can be described in the following formula:

$$B \approx V_{LP} \times t_{ON}$$

The volt-second product  $V_{LP} \times t_{ON}$  defines the magnetization degree of each switching period. If the volt-second products in the above “push” and “pull” periods are not identical, a small direct current component may be generated to cause the deviation of magnetic flow. If balance cannot be restored, the deviation of magnetic flow will gradually increase in the each of the following switching period, making magnetic core become saturated gradually. The phenomenon of the deviation of magnetic flow is usually caused by the unequal on resistance or switching speed of two power switching components. Although the on resistance or switching speed of the two power components are approximately equal through integrating them into the same wafer with the help of high matching advantage of semiconductor integrated circuit technology, the manufacturing error still exists, causing the small deviation for the breakover time.

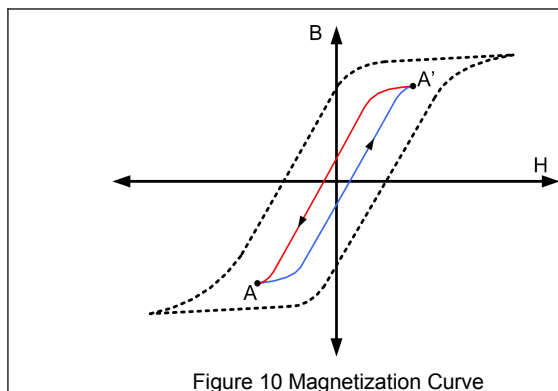


Figure 10 Magnetization Curve

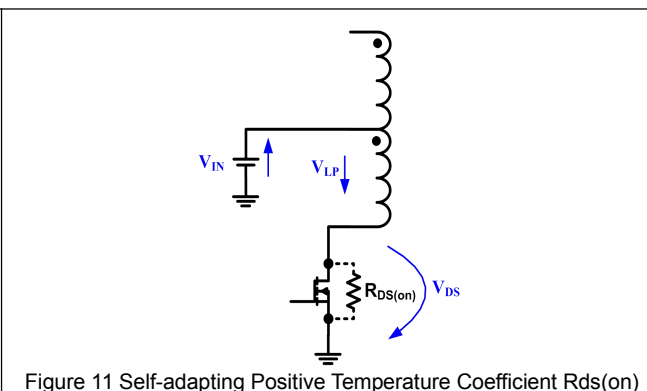


Figure 11 Self-adapting Positive Temperature Coefficient  $R_{DS(on)}$

Fortunately, the breakover resistance  $R_{DS(on)}$  of MOSFET has positive temperature coefficient, the SCM1212B has the self-correcting effect to restrain the imbalance of volt-second based on this feature. Under the condition that there is small deviation between the breakover time of two MOS transistors, the transistor which has longer breakover time  $t_{ON}$  generates more quantity of heat, and the temperature of the transistor rises to improve  $R_{DS(on)}$ , then in the breakover period when the load remains unchanged, the drain-source voltage of this transistor  $V_{DS}$  is relatively high, as shown in Figure 13, the voltage of primary winding  $V_{LP}$  conforms to the formula  $V_{LP} = V_{IN} - V_{DS}$ , thus the  $V_{LP}$  of the transistor which has larger  $t_{ON}$  will gradually decrease to make volt-second recover balance.

SCM1212B has three operation modes, which respectively are start-up mode, operation mode and short-circuit protection mode.

In start-up mode, SCM1212B will first determine whether there is a short circuit phenomenon occurred in the system. The power MOSFET will be driven with limited current during this time. And the chip will choose short setting time.

After eliminating system short circuits, go to overload determination, SCM1212B provides sufficient charging time for output capacitor, to avoid the abnormal start caused by output short circuit which is incorrectly identified due to the excessive low voltage of output capacitor when it is just started, at the same time, the MOS transistor in start-up mode is always operating in current-limiting drive status which means that the drive voltage of MOS transistor in start is limited, thereby making the current flowing through MOS transistor restrained within the safe range of components, that is to restrain the output switching current of primary winding to  $I_{SOFT}$  to achieve the soft start of the system, thereby avoiding the over-current impact and the generation of excessive heat.

In operation mode, MOSFET is always in full drive status which means that the MOS transistor is operating in switching status and the breakover voltage is very low, which guarantees the efficiency of converter.

In short-circuit protection mode, it will stop driving the converter and enters dormancy state. The product will not change to the start-up mode until it dissipates the heat generated in short-circuit mode.

The three operation modes can be freely switched. Only when there is abnormality of output short circuit, the product will repeatedly switch between the start-up mode and short mode, when the abnormality disappears, the product will automatically change to operation mode, all of which can fully guarantee the reliability of converter and have no influence on the performance of converter in normal operation.

### Start-up Mode

When the voltage of the output capacitor is zero when the converter is just started, the converter first enters the short-circuit judgment, and then enters the overload judgment, and the flow schematic is shown in Fig. 12 below, i.e.: start → drive the selected MOS tube with current-limiting drive → detect the MOS tube conduction voltage ( $V_{VD1}$ ,  $V_{VD2}$ ) → judge the relationship between the MOS tube conduction voltage ( $V_{VD1}$ ,  $V_{VD2}$ ) and the set value in accordance with the step-by-step procedure.

#### 1. Step 1 (short-circuit judgment):

If ( $V_{VD1}$ ,  $V_{VD2}$ ) is greater than  $V_{SET1}$ , then accumulate the duration of overvoltage → Judge whether the duration exceeds  $T_{D\_OSP1}$  (28ms, typ.). If the duration does not exceed  $T_{D\_OSP1}$ , then continue to drive the MOS tube in a current-limiting driving mode and detect the relationship between ( $V_{VD1}$ ,  $V_{VD2}$ ) and  $V_{SET1}$  until ( $V_{VD1}$ ,  $V_{VD2}$ ) is detected to be less than  $V_{SET1}$ , then proceed to step 2, i.e., the determination of the MOS tube conduction voltage ( $V_{VD1}$ ,  $V_{VD2}$ ) and  $V_{SET2}$ ; if the duration exceeds  $T_{D\_OSP1}$ , then determine the start-up short-circuit, enter the short-circuit protection mode, stop driving the MOS tube, after a rest time  $T_{SLEEP}$  (1500ms, typ.) to restart the work, and once again enter the start-up mode, continue the cycle;

#### 2. Step 2 (overload determination):

MOS tube conduction voltage ( $V_{VD1}$ ,  $V_{VD2}$ ) and  $V_{SET2}$  determination of two cases:

(1) If ( $V_{VD1}$ ,  $V_{VD2}$ ) is less than  $V_{SET1}$  is greater than  $V_{SET2}$ , then the cumulative time exceeding the duration of  $V_{SET2}$  → determine whether the duration exceeds the  $T_{D\_OSP2}$  (150ms -  $T_{D\_OSP1}$ , typ.). If the duration does not exceed  $T_{D\_OSP2}$ , then continue to drive the MOS tube in current-limit drive mode and detect the relationship between ( $V_{VD1}$ ,  $V_{VD2}$ ) and  $V_{SET2}$  until ( $V_{VD1}$ ,  $V_{VD2}$ ) is detected to be less than the set value of 2, then enter the push-pull steady state mode; if the duration exceeds  $T_{D\_OSP2}$ , then determine the overload, enter the short-circuit protection mode and stop driving the MOS tube. After a rest timing  $T_{SLEEP}$  (1500ms, typ.) to restart the work, once again into the startup mode, continue the cycle;

(2) If ( $V_{VD1}$ ,  $V_{VD2}$ ) is not greater than  $V_{SET2}$ , enter push-pull steady state mode.

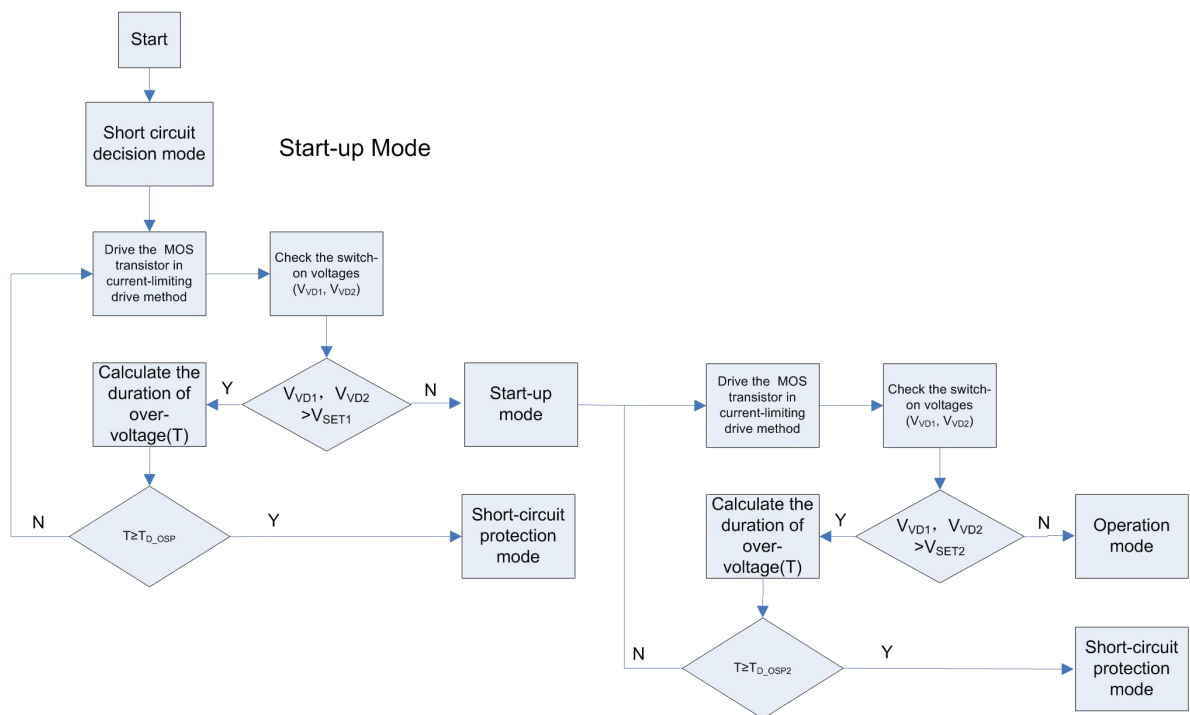


Figure 12 Flow Diagram of Start-up Mode



## Operation Mode

If the output of converter has no short circuit, the voltage of output capacitor, in start-up mode, will gradually increase during the continuously circular charging. When the breakover voltage of MOS transistor is less than or equal to the set value, the converter will turn into operation mode. The flow diagram is shown in Figure 13, that is, determine that the breakover voltage of MOS transistor is less than or equal to the set value → drive the selected MOS transistor with full drive → check the switch-on voltage of MOS transistors → judge whether the voltages ( $V_{VD1}, V_{VD2}$ ) are more than the set value.

If the breakover voltage of MOS transistor is more than the set value, then the system goes into time-counting cycle of start-up mode, otherwise, the system turns back to the step “drive the selected MOS transistor with full drive”, and the above processes are repeated, which is the normal operation of the converter after the product is started. In the operation, MOS transistor is fully driven, that is the MOS transistor is operating in switching status and the on resistance is low, resulting in low energy consumption and high efficiency.

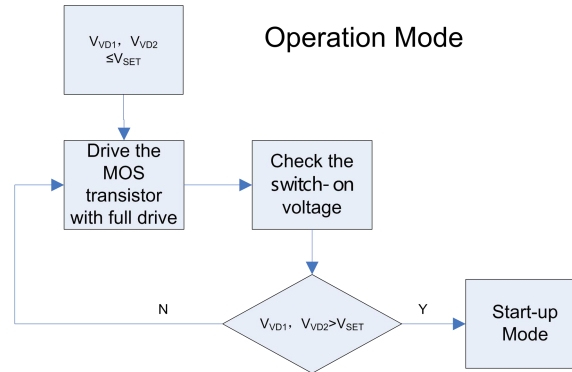


Figure 13 Flow Diagram of Operation Mode

## Short -circuit Protection Mode

If the converter output is in short-circuit state and the on-state voltage of the MOS tube is always detected to be greater than the set value in the start-up mode, there are two cases as follows:

- 1, the on-state voltage of the MOS tube is always greater than  $V_{SET1}$ , then it will accumulate the over-voltage time of 28ms (typ), after which the SCM1212B will stop driving the MOS tube and start calculating the time for stopping driving the MOS tube, and then restart the work after the timing  $T_{SLEEP}$  (1500ms, typ.) to return to the start-up mode.
- 2, The on-state voltage of the MOS tube is always less than  $V_{SET1}$  and more than  $V_{SET2}$ , then it will accumulate the over-voltage time of 150ms (typ.), after which the SCM1212B will stop driving the MOS tube and start calculating the time to stop driving the MOS tube, and then restart the work and return to the start-up mode after the timing  $T_{SLEEP}$  (1500ms, typ.).

The flow diagram of short-circuit protection mode is shown as Figure 14: determine the duration of over-voltage exceeding  $T_{D\_OSP}$  → stop driving and begin to count time (dormancy mode) → finish counting time → turn back to start-up mode. We can see that if the converter is always in output short circuit status, it will operate in the short-circuit protection mode and start-up mode alternately.

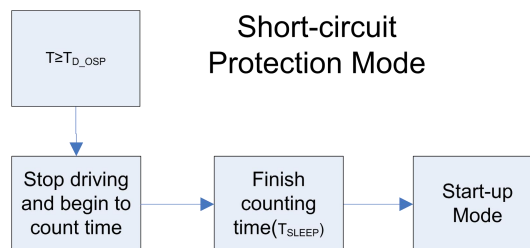


Figure 14 Flow Diagram of Short-circuit Protection Mode

## Extended Output Design

SCM1212B is used to drive the push-pull circuit, which can make output voltage become higher.

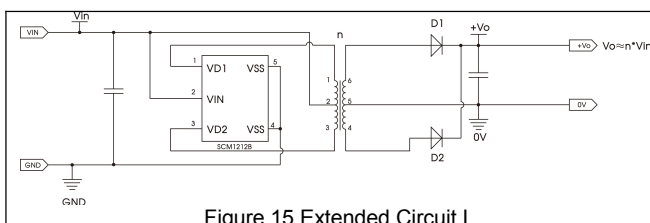


Figure 15 Extended Circuit I

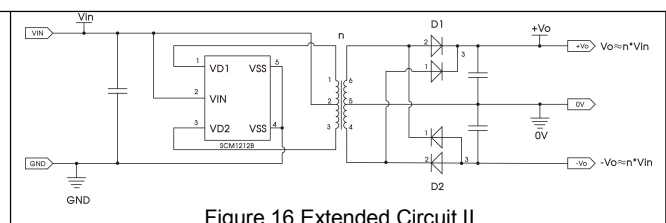


Figure 16 Extended Circuit II



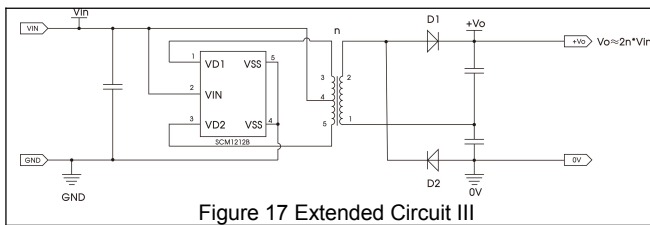


Figure 17 Extended Circuit III

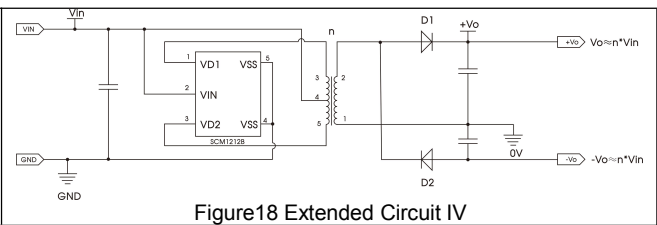


Figure 18 Extended Circuit IV

Extended Circuit I and Extended Circuit II use a full-wave rectification method. The Extended Circuit I is a single output, and the Extended Circuit II is a two-way output. The full-wave rectification topology's side winding is complex, but its output ripple is smaller than the double-pressure rectification.

Extended Circuit III and Extended Circuit IV use a double-pressure rectification method. The Extended Circuit III is a single output, the Extended Circuit IV is a two-way output. The double-pressure rectification topology's side winding is simple, but its output ripple is larger than the full-wave rectification.

## Application Circuit

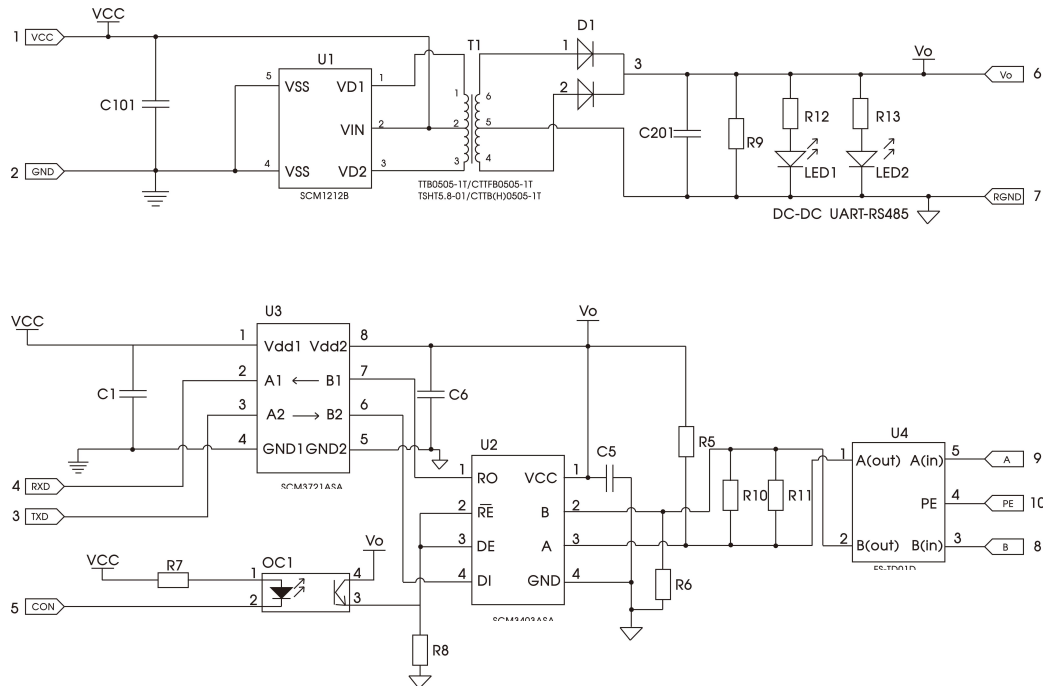


Figure 19 Application Circuit I

### (1) Introduction of TTB05xx-1T Transformer

With the voltage of primary winding and secondary winding of 1650VDC, the allowable operating temperature of  $-40^{\circ}\text{C} \sim +125^{\circ}\text{C}$  and the packaging size of 6.50 x 8.80 x 3.60mm, combined with design of our ICSCM1212BTA product, TTB05xx-1T transformer can be used for electrical isolation scenario which is applicable to 5VDC input and output power less than 1W, such as digital circuit, analog acquisition circuit and data exchange circuit. Please log in the official website of Mornsun and contact the salespeople to obtain the specific specification.

### (2) Introduction of TSHT5.8-01 Transformer

TSHT5.8-01 transformer, with the packaging size of 12.50 x 8.70 x 5.90mm, is specially designed for use with IC. It is mainly used for the electrical isolation scenario which is applicable to 5VDC input and 5VDC output power less than 1W, such as digital circuit, analog acquisition circuit and data exchange circuit.

## Suggested Use of Power Supply

If the input power is not stable enough, it is suggested to add 1uF capacitor in the first section of IC SCM1212B, if there is high requirement to EMI performance, add capacitor and inductor in the first section of the module to filter noise, if there is high requirement to no-load voltage, add resistor after the filtering capacitor of the module as dummy load, it is suggested that the connecting wire of pin 1 and pin 3 to the transformer is as short as possible.

## Ordering information

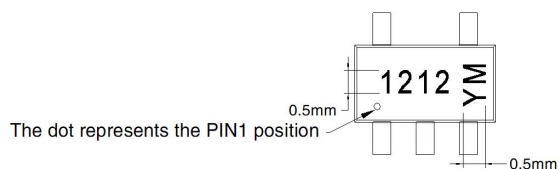
Product Model	Packaging	Quantity of Pin	Silk Screen	Packing
SCM1212BTA	SOT23-5	5	1212YM	3K/tray

### Product model number and screen printing instructions

SCM1212XYZ:

- (1) SCM1212, product code.
- (2) X = A-Z, version code.
- (3) Y = T, packaging code, T: SOT packaging.
- (4) Z = C, I, A, M, code of temperature range: C: 0°C-70°C, I: -40°C-85°C, A: -40°C-125°C, M: -55°C-125°C.
- (5) YM: Date code for product traceability; Y = code for production year; M = code for production month

## Silk Screen Information

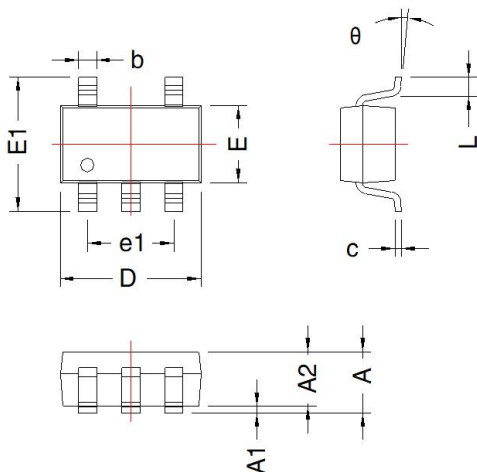


Note:

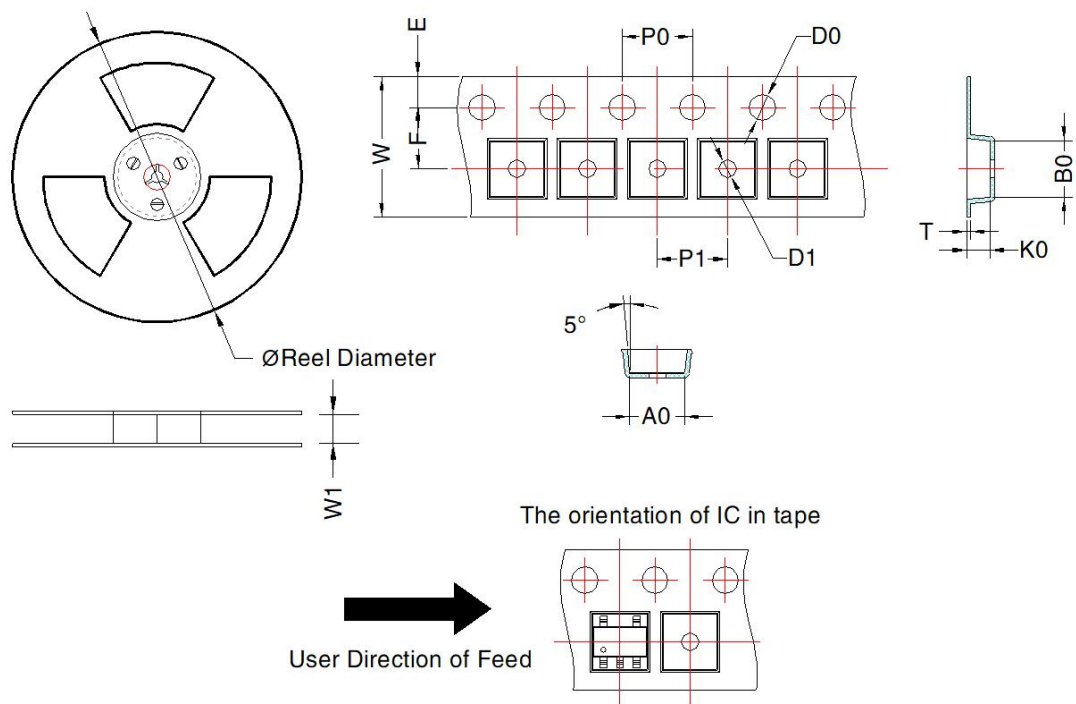
- 1、Typeface: Arial;
- 2、Character size:  
Height: 0.5mm, Spacing: 0.1mm

## Package Information

THIRD ANGLE PROJECTION



SOT-23-5				
Mark	Dimension(mm)		Dimension(inch)	
	Min	Max	Min	Max
A	—	1.30	—	0.051
A1	0	0.1	0	0.004
A2	1.00	1.20	0.040	0.047
D	2.80	3.00	0.110	0.118
E	1.50	1.70	0.059	0.067
E1	2.60	3.00	0.102	0.118
L	0.30	0.60	0.012	0.024
b	0.30	0.50	0.012	0.02
e	0.95 BSC		0.037 BSC	
e1	1.80	2.00	0.071	0.079
c	0.10	0.15	0.004	0.006
θ	0°	8°	0°	8°



Device	Package Type	MPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	T (mm)	W (mm)	E (mm)	F (mm)	P1 (mm)	P0 (mm)	D0 (mm)	D1 (mm)
SCM1212BTA	SOT23-5	3000	180.0	8.5	3.17 ± 0.1	3.23 ± 0.1	1.37 ± 0.1	0.25 ± 0.03	8.0 ± 0.3	1.75 ± 0.1	3.5 ± 0.1	4 ± 0.1	4 ± 0.1	1.5 ± 0.1	1.0 ± 0.1

Note: The minimum order quantity is the minimum packing quantity, and the order quantity shall be an integral multiple of MPQ.

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